

FINITE-TIME STABILIZATION BY USING DEGENERATE FEEDBACK DELAY

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ABSTRACT. Some examples are studied in which a linear controllable dynamical system can be steered towards a specific steady state by using some appropriate linear, time-varying delayed feedback controller. The associated linear retarded differential equation has a finite-dimensional invariant subspace which attracts all orbits in finite time, and this degeneracy property is the reason why the target is attained in finite time rather than just asymptotically.

1. INTRODUCTION

In previous papers [1, 2, 3, 4, 5], Casal, Diaz and the author have considered different variants of delay-differential equations of the type

$$\dot{x} = Ax - M(t)x(t - \tau), \quad t \geq 0, \quad (1.1)$$

where $\tau > 0$ is a given delay, A is the infinitesimal generator of a continuous semigroup on some Banach space X , $M(t)$ is a t -continuous bounded linear map on X whose main characteristic is that it has *compact support* contained on $(0, \infty)$. In this paper we will only deal with the finite-dimensional case, so A and $M(t)$ will be $n \times n$ matrices and $x(t)$ an n -dimensional vector.

Equation (1.1) arises mainly as the *closed-loop* system associated to a general linear, time-invariant controllable system

$$\dot{x} = Ax + Bu(t) \quad (1.2)$$

when a *delayed feedback law* $u(t) = K(t)x(t - \tau)$ is applied for any of the usual purposes of stabilization, tracking, disturbance rejection, etc.

- If $M(t) = 0$ except for $t \in [\tau, 2\tau] \subset (0, \infty)$ then one can prove (see [3]) that every solution $x(t)$ vanishes for $t \geq 2\tau$ if matrix $M(t)$ commutes with e^{At} (or, equivalently, with A), and $\int_{\tau}^{2\tau} M(t)dt = e^{A\tau}$.

- Concerning system (1.2), if $M(t)$ is factorized as $BK(t)$ in order to study the closed-loop delayed feedback system, the situation is much more complicated (see [5]) but similar conclusion can be attained under some quite general circumstances. Also, some optimality properties of the delayed control are studied.

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- These results mean that, even if A is a completely unstable matrix, the closed-loop system is, in fact, *superstable*, that is, all its solutions vanish after some specified finite time. This finite-time exact recovery of a lost equilibrium is usually called *deadbeat control*, (see, e.g., [6]) mostly in the context of the regulator problem. Of course, uniqueness considerations prevent this behavior from happening in “standard” control unless severe discontinuities in the coefficients of the equations are allowed, and it is the fact that the control action starts after a given time lapse is what enables us to handle the problem. In more technical terms, a functional differential equation, which is inherently infinite-dimensional, is *squashed* into \mathbb{R}^n and that fact simplifies many arguments and computations.

- For more general linear systems like $\dot{x} = Ax + z + B(t)u$ (with A nonsingular) the equilibrium $x_{\text{eq}} = -A^{-1}z$ of the uncontrolled system can also be reached in finite time by a similar delayed feedback control. However, its calculation requires previous knowledge of this equilibrium. However, this problem is avoided if, instead of a single-delay feedback $u(t) = K(t)x(t - \tau)$, we try a “Pyragas” type control $u(t) = K(t)[x(t - \tau) - x(t - 2\tau)]$ (see [8, 9]) to eliminate the effect of the nonzero equilibrium.

- As will be shown here, this can also be done for periodic steady states, which is, in fact, more akin to Pyragas’s original purpose (stabilizing unstable periodic orbits in chaotic systems).

2. MAIN RESULT

In this article we consider a more general control system of the type

$$\dot{x} = Ax + f(t) + B(t)u, \quad (2.1)$$

with $f(t)$ a continuous τ -periodic function which corresponds to an external forcing. The associated “unforced system”

$$\dot{x} = Ax + B(t)u, \quad (2.2)$$

will be assumed to be *controllable*.

Theorem 2.1. *Consider the periodically forced linear time-invariant system*

$$\dot{x} = Ax + f(t), \quad (2.3)$$

and assume that the homogeneous system $\dot{x} = Ax$ has no nonconstant τ -periodic solution or, equivalently, the matrix $I - e^{A\tau}$ is invertible. Then this system has a unique τ -periodic solution $p(t)$ and there exists a delayed feedback law

$$u(t) = K(t)[x(t - \tau) - x(t - 2\tau)]$$

of Pyragas type with $K(t) = 0$ outside $[2\tau, 3\tau]$ and such that every solution $x(t)$ of the closed-loop system

$$\dot{x} = Ax + f(t) + B(t)K(t)[x(t - \tau) - x(t - 2\tau)],$$

*is steered toward the **unique** τ -periodic solution $p(t)$ of the forced system (2.1). More specifically*

$$x(t) = p(t) \quad \text{for } t \geq 3\tau.$$

Remark 2.2. The existence and uniqueness of the τ -periodic solution under the hypothesis that $I - e^{A\tau}$ is invertible can be found, for instance, in [7].

Again, irrespective of the possible instability of the matrix A , the (possibly unstable) steady state (a periodic solution in this case) is reached in finite time.

3. THE BASIC CHANGE OF VARIABLES

Let us recall the complete control system

$$\dot{x} = Ax + f(t) + Bu(t), \quad (3.1)$$

where A is $n \times n$, $B(t)$ is $n \times m$ continuous (written simply as B) together with its associated *uncontrolled system* (i.e., for $u(t) = 0$):

$$\dot{x} = Ax + f(t). \quad (3.2)$$

Let us consider new variables $y(t)$ and $w(t)$ related to $x(t)$ by

$$x(t) = e^{At}y(t) + w(t). \quad (3.3)$$

Then $\dot{x}(t) = Ae^{At}y(t) + e^{At}\dot{y}(t) + \dot{w}(t)$ must equal $A[e^{At}y(t) + w(t)] + Bu(t)$ and this gives the “reduced” system

$$\dot{y} = e^{-At}[Aw - \dot{w} + f(t) + Bu(t)]. \quad (3.4)$$

Observe that if $w(t)$ is any solution of the uncontrolled system (3.2), the reduced system (3.4) is just a simple control system $\dot{y} = e^{-At}Bu(t)$ involving only \dot{y} and u but not y .

If $u(t)$ is prescribed to be given by $K(t)[x(t-\tau) - x(t-2\tau)]$, a delayed feedback law, we must also transform this part, thus obtaining the full *transformed closed-loop delay system*

$$\begin{aligned} \dot{y} = e^{-At}[Aw - \dot{w} + f(t)] + e^{-At}BK(t)[e^{A(t-\tau)}y(t-\tau) \\ - e^{A(t-2\tau)}y(t-2\tau) + w(t-\tau) + w(t-2\tau)]. \end{aligned} \quad (3.5)$$

Assume now that $w(t)$ (so far an arbitrary function) is $p(t)$, the unique τ -periodic solution of the uncontrolled system (3.2). Then both $Aw(t) - \dot{w}(t) + f(t)$ and $w(t-\tau) - w(t-2\tau)$ vanish for every t and thus (3.5) becomes

$$\dot{y} = e^{-At}BK(t)[e^{A(t-\tau)}y(t-\tau) - e^{A(t-2\tau)}y(t-2\tau)]. \quad (3.6)$$

First proof of Theorem 2.1. From the previous discussion, if we can show that for some continuous $K(t)$ vanishing outside $[2\tau, 3\tau]$ every solution $y(t)$ of (3.6) becomes zero for $t \geq 3\tau$, then

$$x(t) = e^{At}y(t) + w(t)$$

will be equal to $w(t)$ for $t \geq 3\tau$ as stated in the theorem.

The most direct way to handling this problem is integrating both sides of (3.6) on the interval $[2\tau, 3\tau]$ taking into account that, since $K(t) = 0$ for $t \leq 2\tau$, $y(t)$ is constant on $[0, 2\tau]$ and observing that $y(t-\tau)$ and $y(t-2\tau)$ are equal to $y(0)$ (denoted y_0). Then

$$y(3\tau) = y_0 + \left(\int_{2\tau}^{3\tau} e^{-At}BK(t)e^{At}[e^{-A\tau} - e^{-2A\tau}]dt \right) y_0.$$

Hence $y(3\tau)$ will equal zero for every initial value y_0 if and only if

$$\int_{2\tau}^{3\tau} e^{-At}BK(t)e^{At}dt = [e^{-2A\tau} - e^{-A\tau}]^{-1} = [I - e^{A\tau}]^{-1}e^{2A\tau}, \quad (3.7)$$

where the inverse is well defined as assumed in the statement of the theorem.

Following an argument similar to Sontag's [10], we see that controllability is equivalent to the fact that the "controllability map" $\mathcal{C} : C([2\tau, 3\tau], \mathbb{R}^m) \rightarrow \mathbb{R}^n$ given by

$$\mathcal{C}[u] := e^{3A\tau} \int_{2\tau}^{3\tau} e^{-At} Bu(t) dt$$

is onto, and so is

$$u(\cdot) \mapsto \int_{2\tau}^{3\tau} e^{-At} Bu(t) dt$$

and so is

$$U(\cdot) \mapsto \int_{2\tau}^{3\tau} e^{-At} BU(t) dt$$

defined on the space of continuous matrices $U(t)$ on $[2\tau, 3\tau]$. This means that the matrix Fredholm integral equation

$$\int_{2\tau}^{3\tau} e^{-At} BU(t) dt = [I - e^{A\tau}]^{-1} e^{2A\tau}$$

has at least one solution $\hat{U}(\cdot)$, and setting $K(t) := \hat{U}(t)e^{-At}$ we finally obtain

$$\int_{2\tau}^{3\tau} e^{-At} BK(t)e^{At} dt = [I - e^{A\tau}]^{-1} e^{2A\tau}$$

as desired.

Second proof: optimality considerations. Another possibility is using a well-known explicit expression for a special control $u(t)$ steering any given initial value x_0 to any desired final value x_1 over the time interval $[t_0, t_1]$.

Proposition 3.1 (Minimum-energy control). *Consider the control system $\dot{z} = Pz + Q(t)u(t)$, where P is $n \times n$, and $Q(t)$ is $n \times m$ continuous on $[t_0, t_1]$:*

(1) *The system*

$$\dot{x} = Px + Q(t)u$$

is controllable on $[t_0, t_1]$ if and only if the so-called controllability Gramian

$$W := \int_{t_0}^{t_1} e^{-Pt} Q(t) Q(t)^T e^{-P^T t} dt$$

is nonsingular.

(2) *Assume $\dot{x} = Px + Q(t)u$ is controllable and let $x_0 \in \mathbb{R}^n$. Then: the control law*

$$u^*(t, x_0) := -Q(t)^T e^{-P^T t} W^{-1} x_0$$

minimizes the "total energy"

$$\mathcal{E}[u] := \int_{t_0}^{t_1} u(t)^2 dt$$

over the set of controls steering x_0 to 0 on $[t_0, t_1]$ (which is nonempty by assumption).

For the proof of the above proposition, see, e.g., Sontag [10, Section 3.5]. In our case, the controllable system is just

$$y = e^{-At} B(t)u(t),$$

for which the matrix P above is the zero matrix and $Q(t) = e^{-At}B(t)$. The Gramian is

$$W := \int_{t_0}^{t_1} e^{-At}B(t)B(t)^T e^{-A^T t} dt,$$

which is nonsingular by assumption and the minimum-energy control is $u^*(t, y_0) = U^*(t)y_0$ where $U^*(t)$ is the $n \times n$ matrix

$$U^*(t) = -B(t)^T e^{-A^T t} W^{-1},$$

which “steers the identity matrix I to the zero matrix O on the interval $[t_0, t_1]$ ”.

In our case, we need $u^*(t, y_0)$ to be of the delayed feedback type

$$u^*(t, y_0) = K(t)[e^{A(t-\tau)}y(t-\tau) - e^{A(t-2\tau)}y(t-2\tau)].$$

As previously pointed out, $y(t-\tau) = y(t-2\tau) = y_0 = y(0)$. Therefore, we must find an $m \times n$ matrix $K^*(t)$, vanishing at the endpoints 2τ and 3τ , such that

$$-B(t)^T e^{-A^T t} W^{-1} = K^*(t)e^{At}(e^{-A\tau} - e^{-2A\tau}) = K,$$

which is just

$$K^*(t) = -B(t)^T e^{-A^T t} W^{-1}(e^{-A\tau} - e^{-2A\tau})^{-1}. \quad (3.8)$$

The feedback law is thus obtained by extending this K^* outside of $[2\tau, 3\tau]$ by setting $K^*(t) = 0$.

4. FINAL REMARKS

(1) Some of the above results work perfectly well for linear, time-varying systems $\dot{x} = A(t)x + B(t)u$ by substituting e^{At} , e^{-As} by $\Phi(t)$, $\Phi(s)^{-1}$, where $\Phi(t)$ is the fundamental matrix solution satisfying $\Phi(0) = I$. The existence of a unique periodic solution to $\dot{x} = A(t)x + f(t)$ for τ -periodic $f(t)$ is, of course, quite a difficult matter, even requiring the coefficient matrix $A(t)$ to be also τ -periodic (see [7]).

(2) From a practical viewpoint, the question of *robustness* presents itself immediately. Exact equilibria attained in finite time, deadbeat control, etc., are not found in real life, since neither the plants nor the control links are 100% valid. If all the eigenvalues of matrix A have negative real parts, the consequences are not so bad, since the system’s own internal dynamics will drive the state back to (a neighborhood of) equilibrium. But in the unstable case, this will not be true.

A possible solution to this problem is to extend the control to a stream of equal actions on intervals $[2\tau, 3\tau]$, $[5\tau, 6\tau]$, $[8\tau, 9\tau]$, and so on. The “inaction intervals” of length 2τ enable the system to “forget past history” and start all over again. No recovery of equilibrium (or some other periodic steady state) will happen, but at least we will make sure that the system will not deviate too far from it. This idea comes quite close to the so-called “act-and-wait” control strategy or “intermittent control” (see [6]).

(3) It is well-known that sparse jump discontinuities in controller functions are not a real problem (at least in finite-dimensional problems), and our delayed feedback gain function $K(t)$ is usually discontinuous at 2τ and 3τ . Yet, if they are considered undesirable in some specific situation, these jumps can easily be avoided by choosing a continuous scalar function $\beta : \mathbb{R} \rightarrow \mathbb{R}$ such that $\beta(t) > 0$ on $[2\tau, 3\tau]$ and is 0 outside this interval, and modifying our original system $\dot{x} = Ax + B(t)u(t)$ by $\dot{x} = Ax + \beta(t)B(t)u(t)$. If the former system is controllable on $[2\tau, 3\tau]$, the latter will have the same property, as can be easily proven (the assumption “ $\beta > 0$ ”

on $[2\tau, 3\tau]$ plays an important role). We then pick the new $K(t)$ as in (3.8) with $B(t)^T$ substituted by $\beta(t)B(t)^T$ and W redefined as

$$\int_{2\tau}^{3\tau} \beta(t)^2 e^{-At} B(t) B(t)^T e^{-A^T t} dt :$$

$$K_{\text{new}}^*(t) := -\beta(t)B(t)^T e^{-A^T t} W^{-1} (e^{-A\tau} - e^{-2A\tau})^{-1}.$$

Some interesting questions arise as to the “right” choice of $\beta(t)$, depending on the performance index associated to the problem under study. Some results in this line will appear elsewhere.

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